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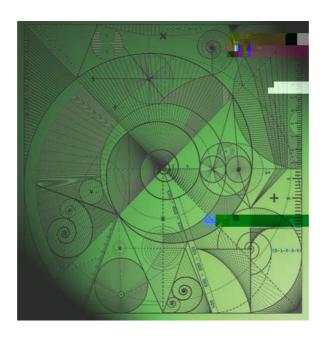
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Regularization of Descriptor Systems

by

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Abstract Implicit dynamic-algebraic equations, known in control theory as descriptor systems, arise naturally in many applications. Such systems may not be regular (often referred to as singular). In that case the equations may not have unique solutions for consistent initial conditions and arbitrary inputs and the system may not be controllable or observable. Many control systems can be regularized by proportional and/or derivative feedback. We present an overview of mathematical theory and numerical techniques for regularizing descriptor systems using feedback controls. The aim is to provide stable numerical techniques for analyzing and constructing regular control and state estimation systems and for ensuring that these systems are robust. State and output feedback designs for regularizing linear time-invariant systems are described, including methods for disturbance decoupling and mixed output problems. Extensions of these techniques to time-varying linear and nonlinear systems are discussed in the final section.

1 Introduction

Singular systems of differential equations, known in control theory as *descriptor systems* or *generalized state-space systems*, have fascinated Volker Mehrmann throughout his career. His early research, starting with his habilitation [33, 35],

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where $E,A\in\mathbb{R}^{n\times n},B\in\mathbb{R}^{n\times m},C\in\mathbb{R}^{p\times n}$. Here $x(\cdot)$ is the state, $y(\cdot)$ is the output, and $u(\cdot)$ is the input or control of the system. It is assumed that $m,p\leq n$ and that the matrices B,C are of full rank. The matrix E may be singular. Such systems are known as descriptor or generalized state-space systems. In the case E=I, the identity matrix, we refer to (1) or (2) as a standard system.

We assume initially that the system is time-invariant; that is, the system matrices E,A,B,C are constant, independent of time. In this context, we are interested in proportional and derivative feedback control of the form $u(t) = Fy(t) - G\dot{y}(t) + v(t)$ or u(k) = Fy(k)

$$\operatorname{rank}(\begin{bmatrix} E+BG \\ T_{\infty}^T(E+BG)(A+BF) \end{bmatrix})=n. \tag{13}$$

$$(E+GC,A+FC), (15)$$

where the matrice \mathcal{E} and \mathcal{G} must be selected to ensure that the response the observer converges to the system state \mathcal{E} any arbitrary starting condition; that is, the system must be asymptotically stable. By duality white state feedback problem, it follows that if the condition \mathcal{E} holds, then the matrice and \mathcal{E} can be chosen such that the corresponding closed-loop pencils 160 ular and of index at most one. If condition 160 also holds, then the closed-loop system is 160 ensure the stability and robustness of the system and the nite eighth 160 ensure the stability and robustness of the system and the nite eighth 160 ensure the sassigned explicitly by the techniques described 160 ensure feedback control problem.

3.2 Disturbance Decoupling by State Feedback

In practice control systems are subject to disturbances tha

$$UEV = t_{3} \begin{cases} 2 & t_{1} & t_{2} & t_{3} & s_{4} & s_{5} \\ E_{11} & 0 & 0 & 0 & 0 \\ t_{2} & E_{21} & E_{22} & 0 & 0 & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{24} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & 0 \\ E_{21} & E_{22} & E_{24} & E_{24} & E_{24} \\ E_{21} & E_{22} & E_{23} & E_{33} & E_{34} & 0 \\ E_{21} & E_{22} & E_{23} & E_{34} & E_{34} \\ E_{21} & E_{22} & E_{23} & E_{34} & E_{34} \\ E_{21} & E_{22} & E_{33} & E_{34} & 0 \\ E_{21} & E_{22} & E_{33} &$$

WCV =

$$E + BG\Gamma = {M + B_1G \ 0 \over K + B_2G \ I} , A + BFC = {0 \ I + B_1F \over P \ B_2F} .$$
 (26)

Different effects can, therefore, be achieved by feeding the the derivatives or the states. In particular, in the case where is singular, but $ran[M,B_1]=n$, the feedback can be chosen such that $+B_1G$ is invertible and well-conditioned [7], giving a *robust* closed-loop system that is regular and of index zero. The track matrix F can be chosen separately to assign the eigenvalues of the robust [30], for example, or to achieve other objectives.

The complete solution to the mixed output feedback regadation problem is given in [22]. The theorem and its proof are very technical bility is established using condensed forms derived in the paper. The solution to the coutput feedback problem given in Theorem 4 is a special case of the complete the remixed output case given in [22]. The required feedback matrices anstructed directly from the condensed forms using numerically stable transations.

Usually the design of the feedback matrices still containeedom, however, which can be resolved in many different ways. One choice is the feedbacks such that the closed-loop system is robust, or insensitive turbations, and, in particular, such that it remains regular and of index at most under perturbations (due, for example, to disturbances or parameter varial tions choice can also be shown to maximize a lower bound on the stability radius of the ed-loop system [13]. Another natural choice would be to use minimum noticed backs, which would be a least squares approach based on the theory in [24] approach is also

$$E(t)x(t) = A(t)x(t) + B(t)u(t), \quad x(t_0) = x_0,$$

$$y(t) = C(t)x(t),$$
 (27)

where $E(t), A(t) \in \mathbb{R}^{n \times n}, B(t) \in \mathbb{R}^{n \times m}, C(t) \in \mathbb{R}^{p \times n}$ are all continuous functions of time and x(t) is the state y(t) is the output, and y(t) is the input or control of the system. (Corresponding discrete-time systems with tians y(t) coef cients can also be defined, but these are not considered here.)

In this general form, complex dynamical systems includingstraints can be modelled. Such systems arise, in particular, as line award a general nonlinear control system of the form

F
$$(t,x,x,u) = 0$$
, $x(t_0) = x_0$, $y = G(t,x)$, (28)

where the linearized system is such $t\mathbb{E}(t)$, A(t), B(t) are given by the Jacobians of F with respect to x, x, u, respectively, and C(t) is given by the Jacobian $d\mathbb{E}(t)$ with respect to x (see [31]).

For the time-varying system (27) and the nonlinear system), (the system properties can be modiled by time-varying state and output feachtbas in the time-invariant case, but the characterization of the system aintiqual ar the solvability and regularity of the system, is considerably more competitate denote than in the time-invariant case and it is correspondingly more difficult analyse the feedback

E; A; B; C; are assumed to be analytic functionstofbut these conditions can be relaxed provided the ASVD decompositions remain suf diestmooth.

In the papers [12, 31], a much deeper analysis of the regulation problem is developed. Detailed solvability conditions for the tirregrying system (27) are established and different condensed forms are derived, aging the ASVD. Constant rank assumptions do not need to be applied, althoughvistence of smooth ASVDs are required. The analysis covers a plethora of differpossible behaviours of the system. One of the tasks of the analysis is to determinendancies and inconsistencies in the system in order that these may be exclitred the design process. The reduction to the condensed forms displaysealint ariants that determine the existence and uniqueness of the solution. The index system is then de ned to be regularizable if there exist proportional or index for consistent initial state vector and any given (ilos08(i).46tnnsmooth.sy eerensyho

5 Conclusions

We have given here a broad-brush survey of the work of VolkehMhann on the problems of regularizing descriptor systems. The <code>extecth</code> is work alone is formidable and forms only part of his research during hiseearWe have concentrated speci cally on results from Volker's own approaches the regularity problem. The primary aim of his work has been to provide stable entirel techniques for analyzing and constructing control and state estimations and for ensuring that these systems s1ns1nwsons. The s

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